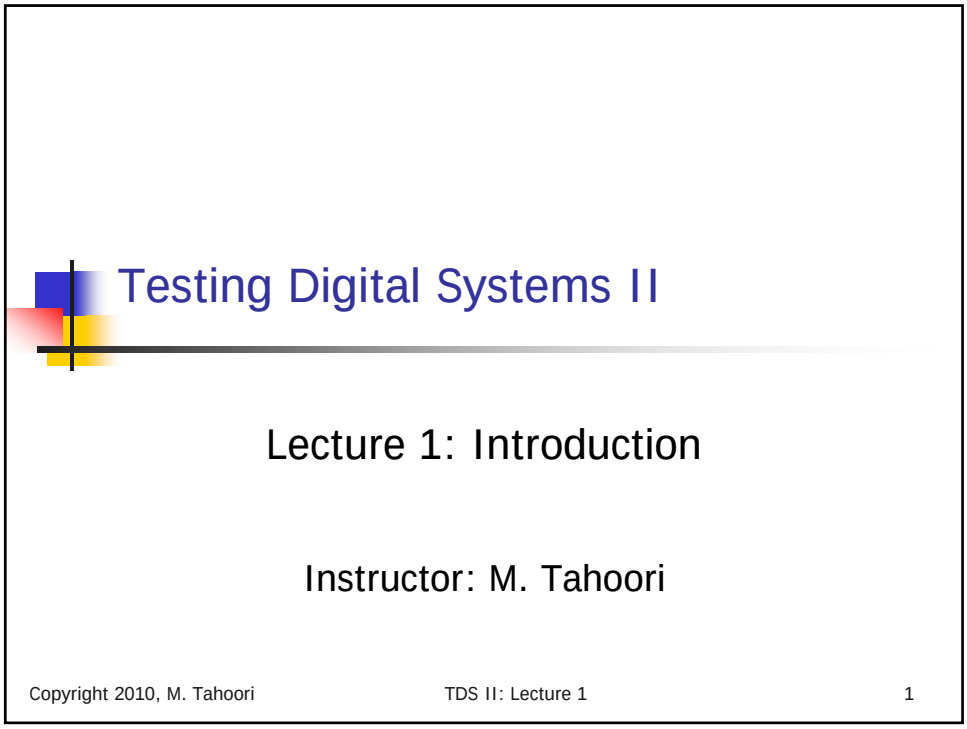
Testing Digital Systems II
Lecture 1: Introduction
Instructor: M. Tahoori
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TDS II: Lecture 1
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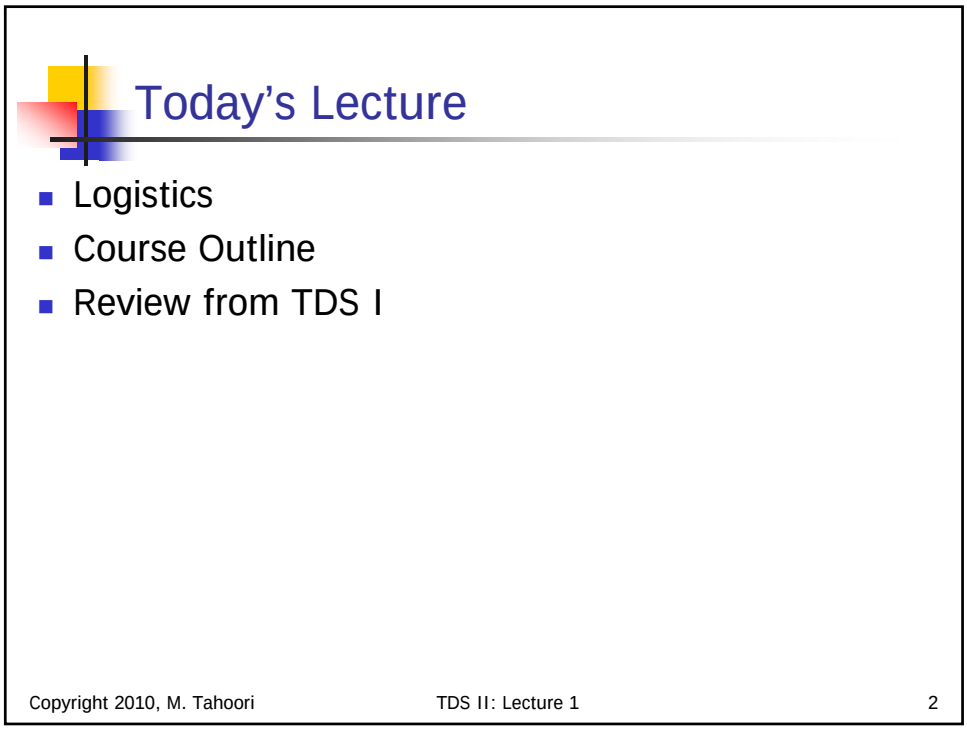
Today's Lecture
Logistics
Course Outline
Review from TDS I
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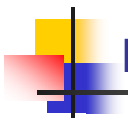
Logistics

- Instructor: Mehdi Tahoori
  - Office: Room B2-313.1, Building 07.21
  - Email: mehdi.tahoori@kit.edu
  - Tel: 721-608-7778, Fax: 721-608-3962
- Lecture:
  - When: Wednesdays 14:00-15:30
  - Where: Room 301, Building 50.34

Logistics (cont)

- Requirements
  - Logic Design
  - Computer Architecture
  - Testing Digital Systems I
  - Background on
    - Algorithms and Programming
    - Hardware description languages (VHDL or Verilog)



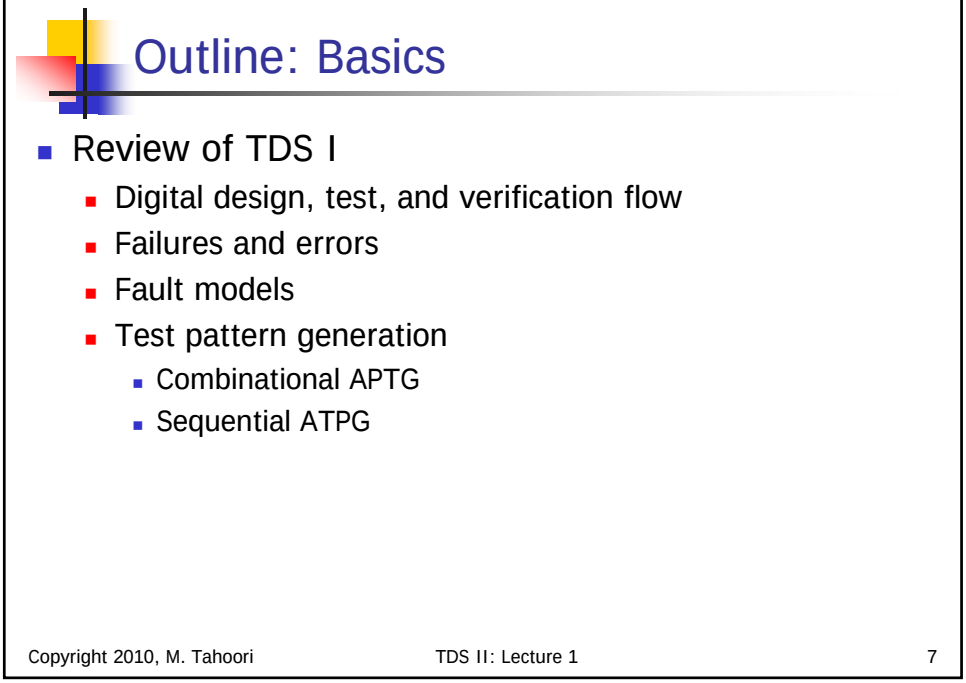
Reference Books

- Textbook
  - **Essentials of Electronic Testing for Digital, Memory and Mixed-Signal VLSI Circuits** by M. L. Bushnell and V.D. Agrawal, *Kluwer Academic Press*, Boston 2000
- Recommended
  - **System On Chip Test Architectures: Nanometer Design for Testability** by L.T. Wang, C.E. Stroud, N. A. Touba, Elsevier, Morgan Kaufmann Publishers, 2009.
  - **Digital System Testing and Testable Design** by M. Abramovici, M. A. Breuer, and A.D. Friedman, *IEEE Press*, New York, 1990, 652 pages

Course Outline

- Basics
- Design for Testability (DFT)
- Built-in Self-test (BIST)
- More topics on test generation and DFT

- I try to be flexible. The order and contents may be changed as we proceed.


Outline: Basics
Review of TDS I
Digital design, test, and verification flow
Failures and errors
Fault models
Test pattern generation
Combinational APTG
Sequential ATPG
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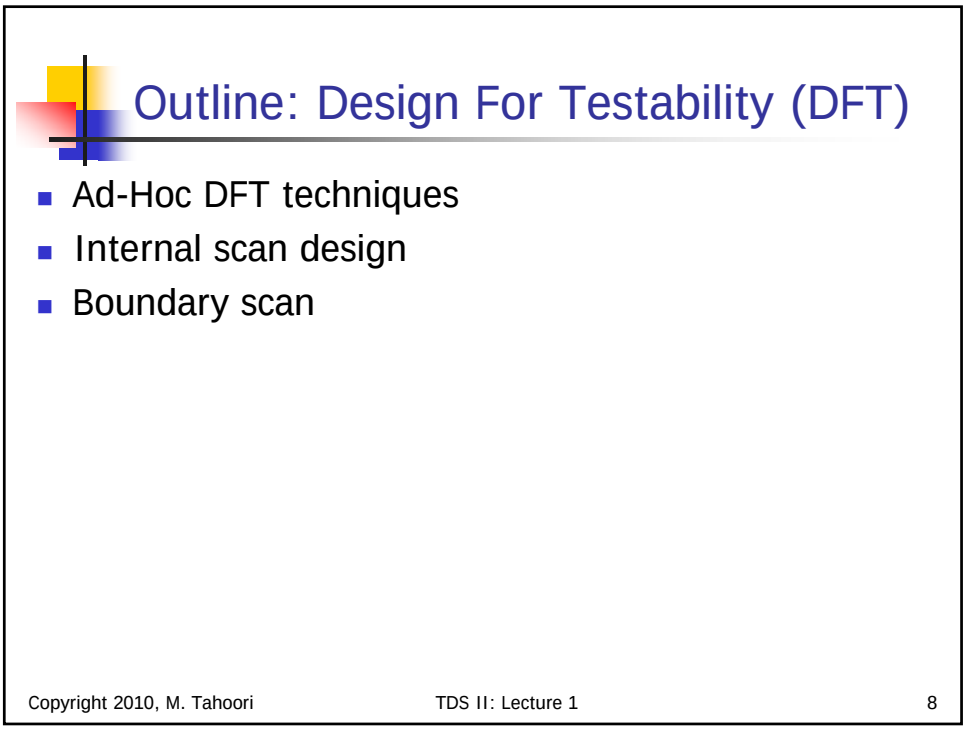
Outline: Design For Testability (DFT)
Ad-Hoc DFT techniques
Internal scan design
Boundary scan
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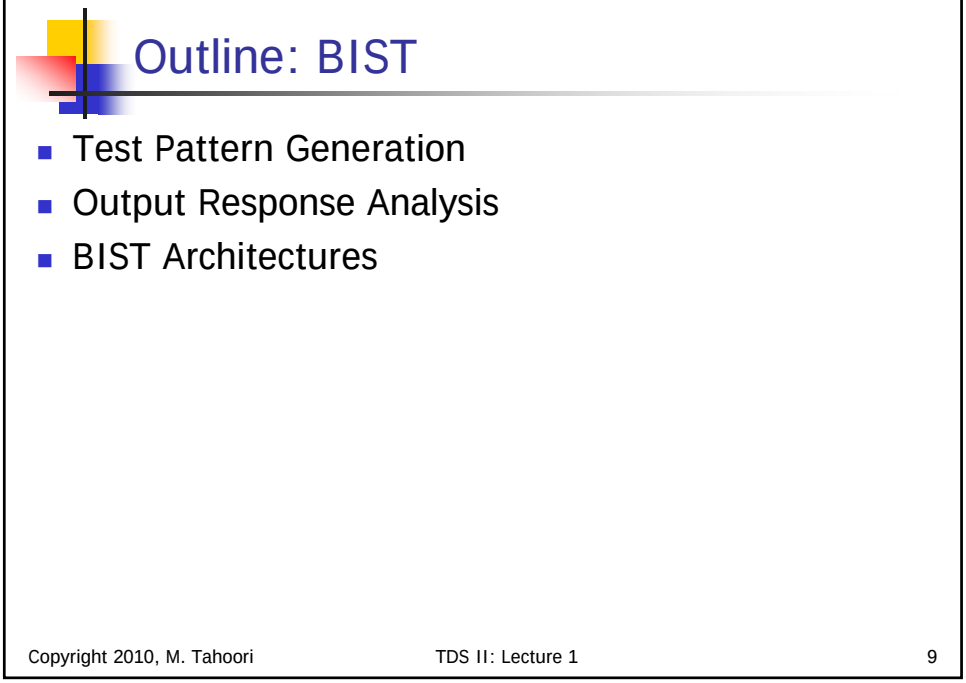
Outline: BIST
Test Pattern Generation
Output Response Analysis
BIST Architectures
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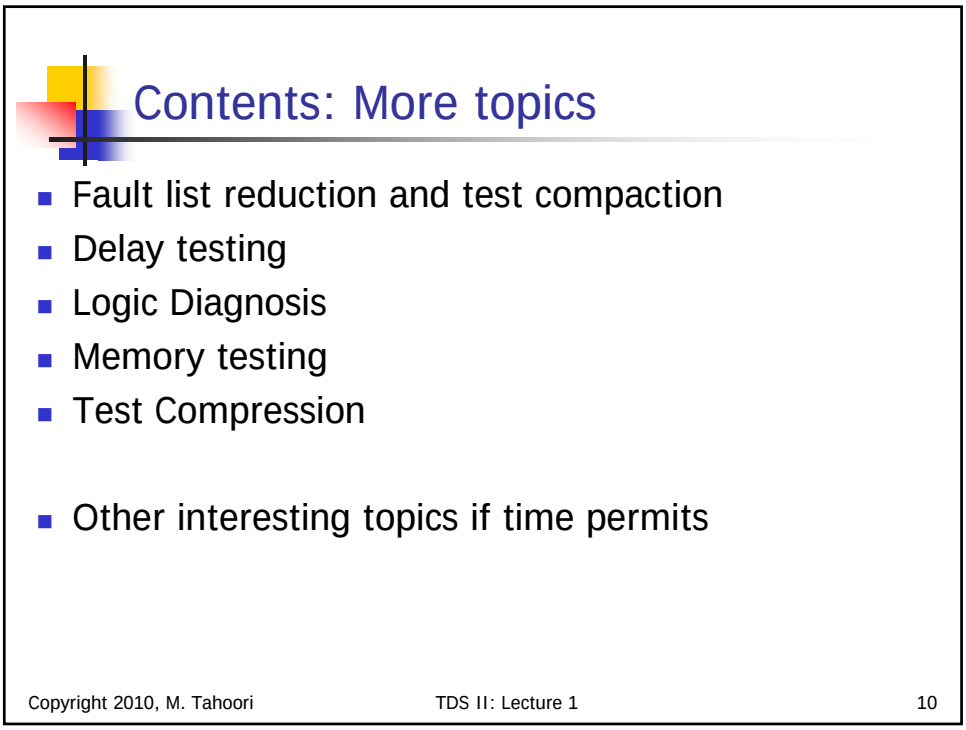
Contents: More topics
Fault list reduction and test compaction
Delay testing
Logic Diagnosis
Memory testing
Test Compression
Other interesting topics if time permits
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Overview of TDS I

VLSI Realization Process

Customer's need

Determine requirements

specifications

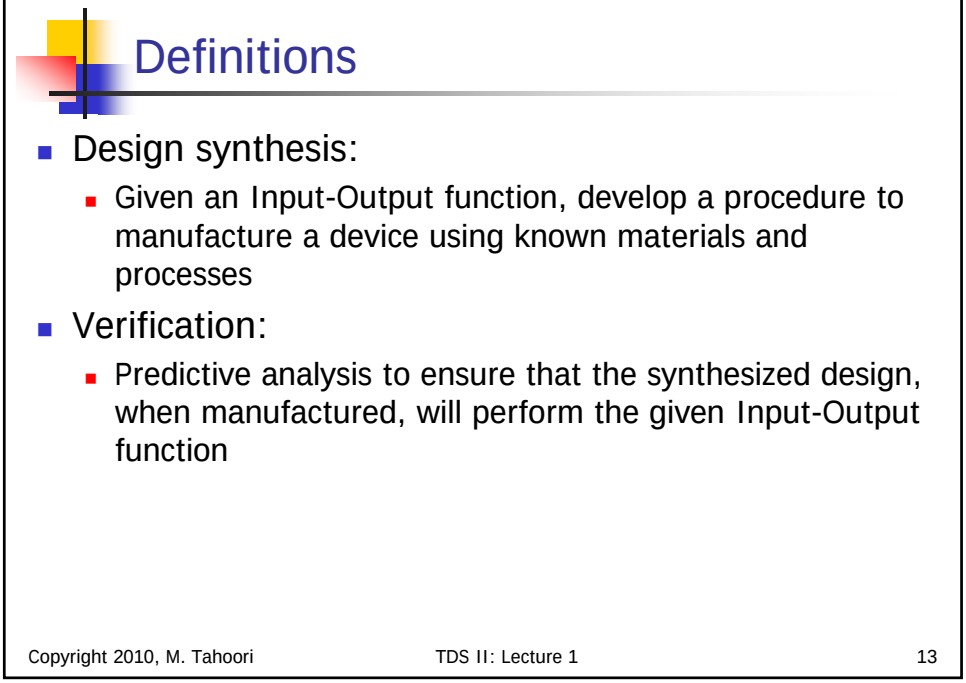
Design synthesis and Verification

Test development

Fabrication

Manufacturing test

Chips to customer


Definitions
Design synthesis:
Given an Input-Output function, develop a procedure to manufacture a device using known materials and processes
Verification:
Predictive analysis to ensure that the synthesized design, when manufactured, will perform the given Input-Output function
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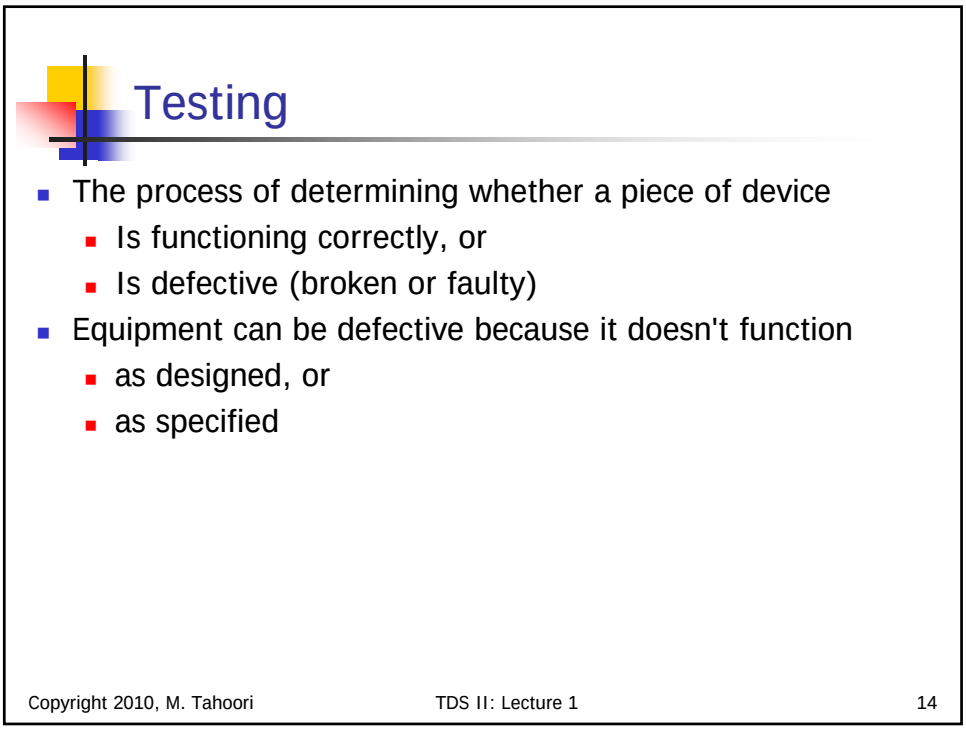
Testing
The process of determining whether a piece of device
Is functioning correctly, or
Is defective (broken or faulty)
Equipment can be defective because it doesn't function
as designed, or
as specified
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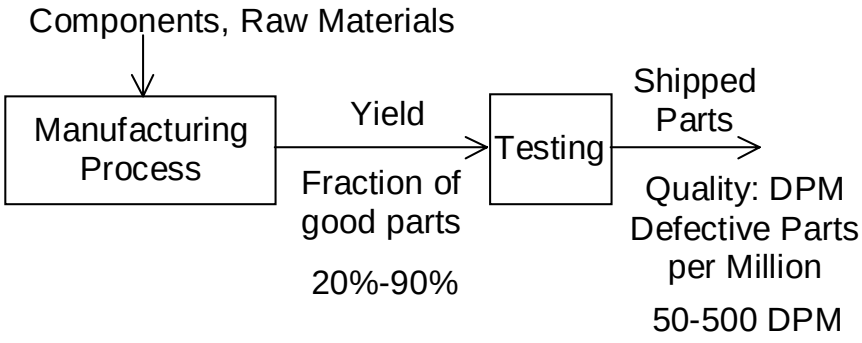
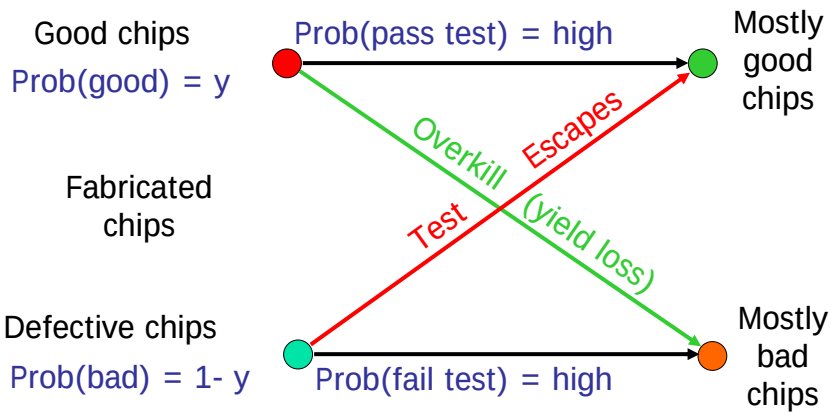
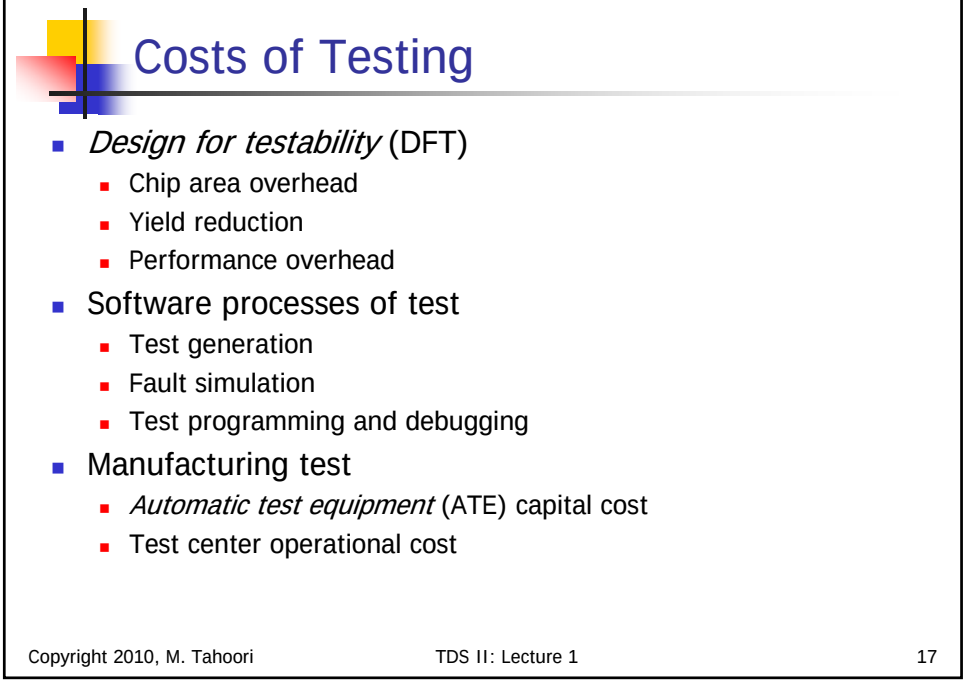
Testing Process



Testing as Filter Process




Costs of Testing
Design for testability (DFT)
Chip area overhead
Yield reduction
Performance overhead
Software processes of test
Test generation
Fault simulation
Test programming and debugging
Manufacturing test
Automatic test equipment (ATE) capital cost
Test center operational cost
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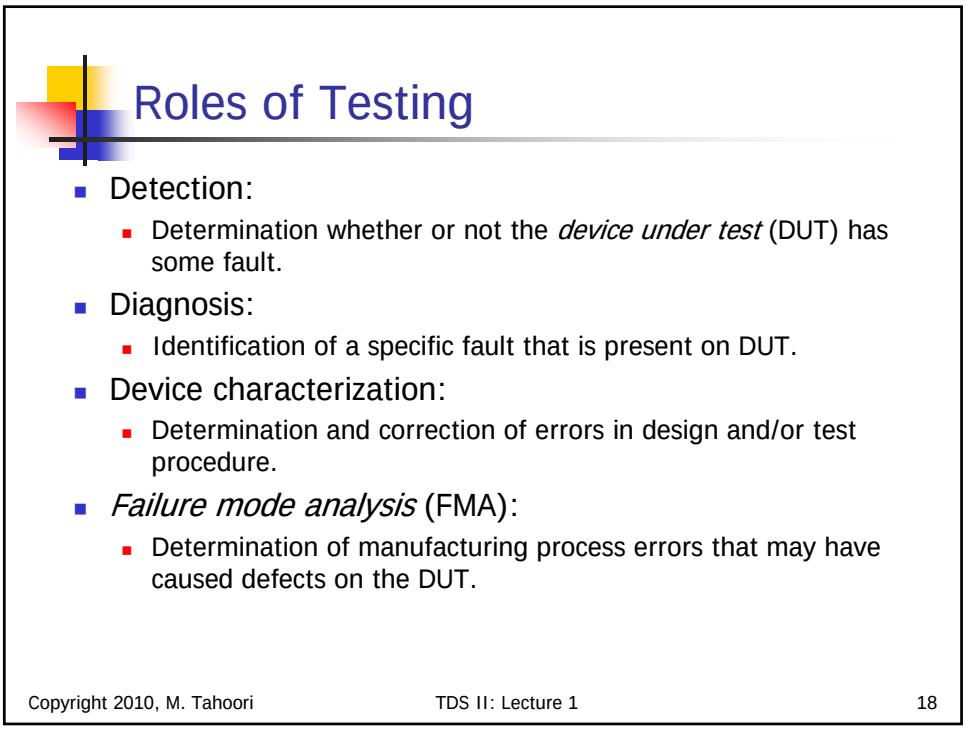
Roles of Testing
Detection:
Determination whether or not the device under test (DUT) has some fault.
Diagnosis:
Identification of a specific fault that is present on DUT.
Device characterization:
Determination and correction of errors in design and/or test procedure.
Failure mode analysis (FMA):
Determination of manufacturing process errors that may have caused defects on the DUT.
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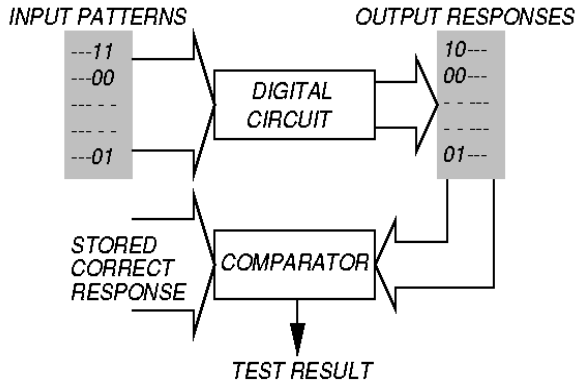
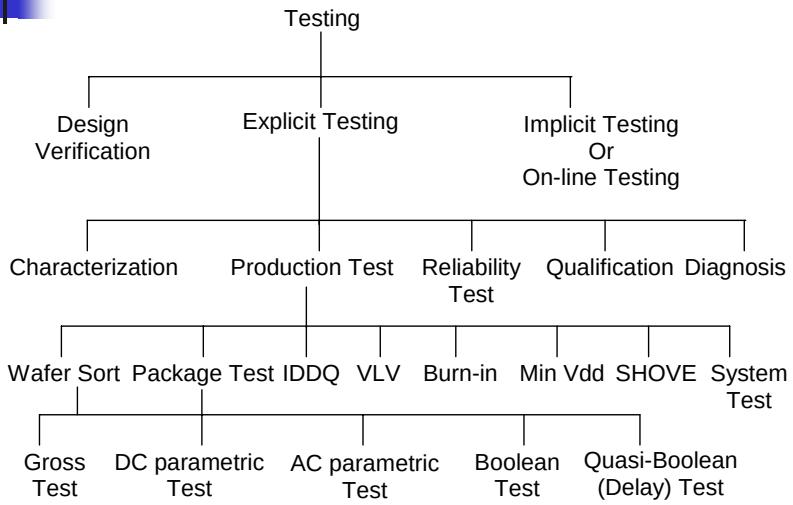
Testing Principle

- Performed by
  - Automatic Test Equipment (ATE)
  - On-chip Built-in Self Test (BIST)



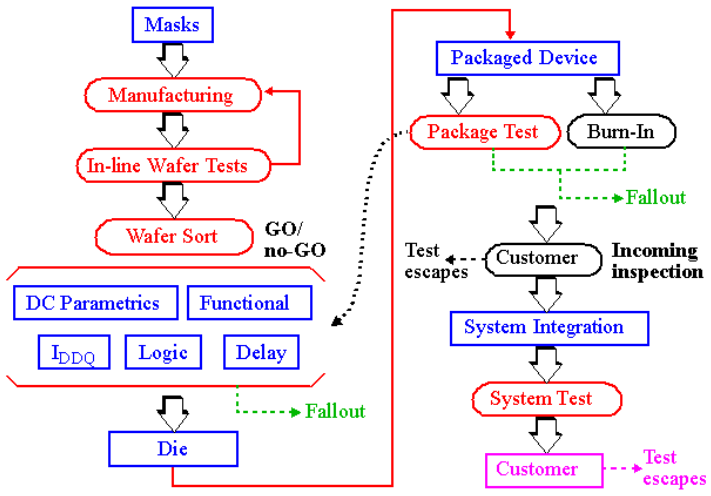
Testing Taxonomy



Types of Test

Production Test	Tests to sort out defective manufactured parts
Wafer Sort or Probe	Test of each die while still on the wafer
Final or Package Test	Test of packaged chips and separation into classes or bins (military, commercial, industrial)
Acceptance Test	A test to demonstrate the degree of compliance of a device with purchaser's requirements
Sample Test	Test of some but not all parts
Go / No Go Test	Test to determine whether device meets specification
Characterization	Test to determine actual values of device AC and DC parameters and the interaction of parameters. Used to set final specifications and to identify areas to improve process to increase yield.
Stress Screening	Test with stress (high temperature, temperature cycling, voltage, vibration, etc.) applied to eliminate short life parts
Reliability Test (Accelerated Life Test)	Test after subjecting the part to extended high temperature or voltage to estimate time to failure in normal operation
Diagnostic Test	Test to locate failure site on failed part
Quality Test	Test by quality assurance department of a sample of each lot of manufactured parts. More stringent than final test.
On-line Test*	On-line testing to detect errors that occur during normal system operation.
System Test	Test by plugging a device into an actual system and running the system.
Design Verification	Verifying the correctness of a design

Test Flow



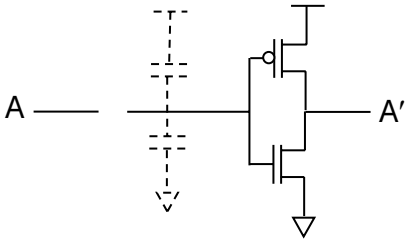
Failures, Errors, Faults

- Chip is Defective if
  - it Doesn't Function
    - as Specified, or
    - as Designed due to Presence of a Failure
- Error
  - Incorrect Signal Value
- Failure
  - Deviation from Designed Characteristics
- Fault
  - Models Effect of Failure on Logical Signals

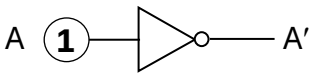
Open

Failure Mechanism: Particle on

Failure Mode: Open Metal



Electrical Fault: Open Signal

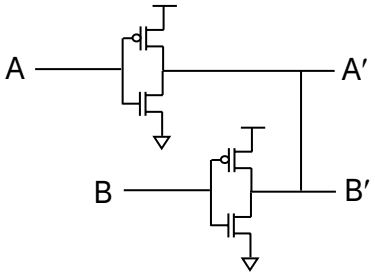


Logical Fault Model: stuck-at-1

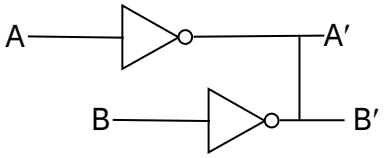
Short

Failure Mechanism: Particle on

Failure Mode: Shorted Metal



Electrical Fault

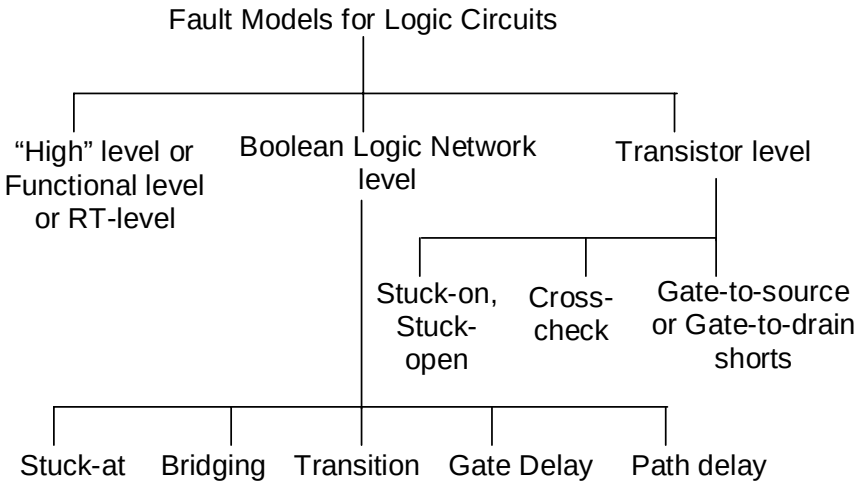


Logical Fault

Fault Model

- Fault model
  - Models effect of physical failure on logic network
    - Abstraction of physical situation
    - Used to describe the change in the logic function of a device caused by the defect.
- Various levels of abstraction are used
  - Functional (Board, Chip) level
  - Register transfer (Behavioral) level
  - Logic level
    - Gate library level
    - Elementary gate level
  - Switch level
  - Transistor (Spice) level

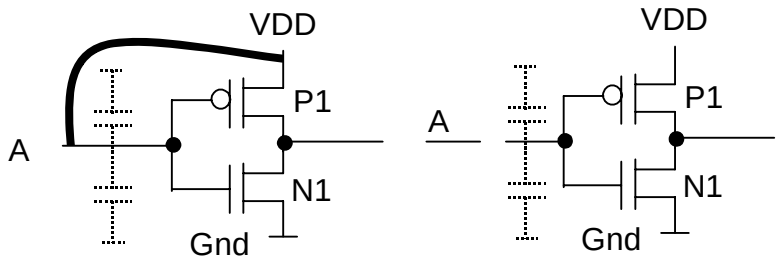
Fault Model Taxonomy



Stuck Fault Models

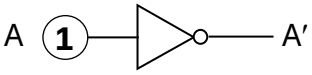
- Structural logic-level fault model
  - Start with the circuit represented as a netlist of Boolean gates
  - Assumes faults only affect the interconnection between gates
- Single Stuck Fault
  - Logic network of elementary gates
    - AND, OR, NAND, NOR, NOT
  - One Line has Fixed 0 or 1 Value
  - Independent of other signal values
  - One fanout branch can be stuck
  - Most common model for Boolean test
  - Written L_i / h, h = 0 or 1
- Multiple Stuck Fault
  - One or More Stuck Line Faults Present
- Pin Fault
  - Stuck Fault on I/O Connection of a Module

Stuck-at Fault Model



Short to VDD

Open signal lead



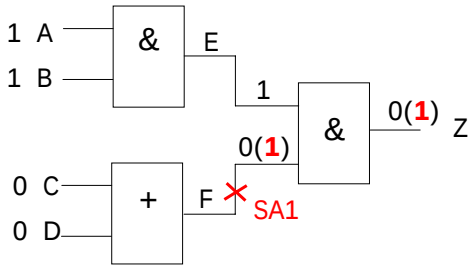
Notation: **A/1** or $\mathbf{A_1}$

Fault Detection

- An input combination *detects* a fault in a logic network if
  - the response of the faulty logic network to that input combination is different from that of the fault-free network
  - The input combination is called a *test pattern* for the fault
- Fault detection requires:
  - A test t activates or provokes the fault f.
  - t propagates the error to an observation point
    - e.g. primary output
- A line whose value changes with f present is said to be sensitized to the fault site.

Single Stuck-at

- 14 faults
  - 2 faults (SA0, SA1) per each line
- ABCD = 1100 detects F/1
  - Faulty and fault-free outputs different
- ABCD = 1101 does NOT detect F/1
  - Faulty and fault-free outputs are the same



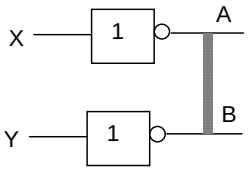
Untestable Fault

- A fault that does not affect the logical behavior of a circuit (redundant fault)
  - Untestable by Particular Test Procedure
- Causes
  - Redundant Circuitry
    - Design Error
  - Hazard Control Circuitry
  - Error Detection Circuitry
    - Parity Check
  - Excess Components
    - Needed for Performance, not Functionality

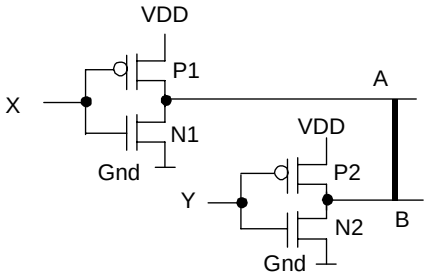
Bridging Fault

- *Bridging faults* appear when two or more normally distinct signal lines in a Boolean logic network are unintentionally shorted together and create wired logic.
- A *feedback bridging fault* is a special type of bridging fault which is created when one of the two shorted signal lines depends on the other signal line in the fault-free circuit.
  - May cause oscillation or latch
- If a fanout branch of a signal line is involved in a bridge
  - Logic value on the fanout stem and the other fanout branches of that signal line will be the same as the logic value on the fanout branch which is involved in the bridge

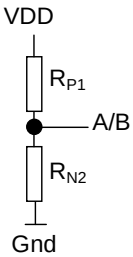
Bridging Fault



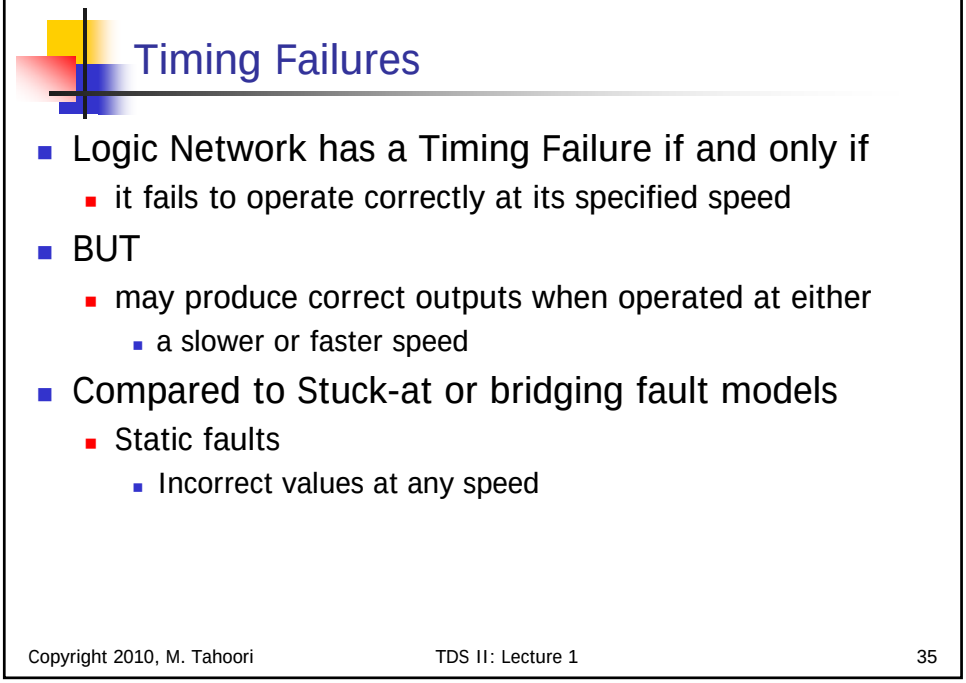
Logic-level model



Transistor-level model

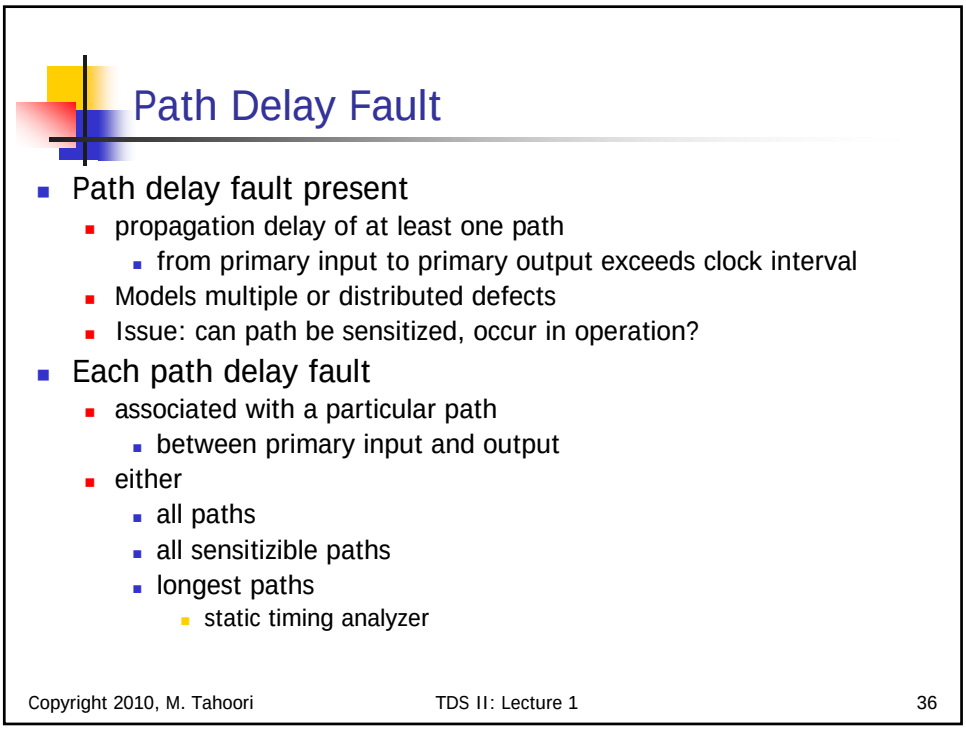


Electrical model when X = 0 and Y = 1



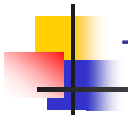
Timing Failures

- Logic Network has a Timing Failure if and only if
  - it fails to operate correctly at its specified speed
- BUT
  - may produce correct outputs when operated at either
    - a slower or faster speed
- Compared to Stuck-at or bridging fault models
  - Static faults
    - Incorrect values at any speed



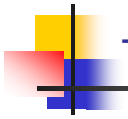
Path Delay Fault

- Path delay fault present
  - propagation delay of at least one path
    - from primary input to primary output exceeds clock interval
  - Models multiple or distributed defects
  - Issue: can path be sensitized, occur in operation?
- Each path delay fault
  - associated with a particular path
    - between primary input and output
  - either
    - all paths
    - all sensitizible paths
    - longest paths
      - static timing analyzer



Transition Fault

- Definition
  - A localized timing failure is large enough such that the delay of all paths through some gate to observable outputs exceed the clock interval
- Each transition fault
  - associated with a particular gate input or gate output
  - either a 0 to 1 transition or a 1 to 0 transition
    - (two transition faults)
      - slow-to-rise, slow-to-fall
  - propagated to some primary output



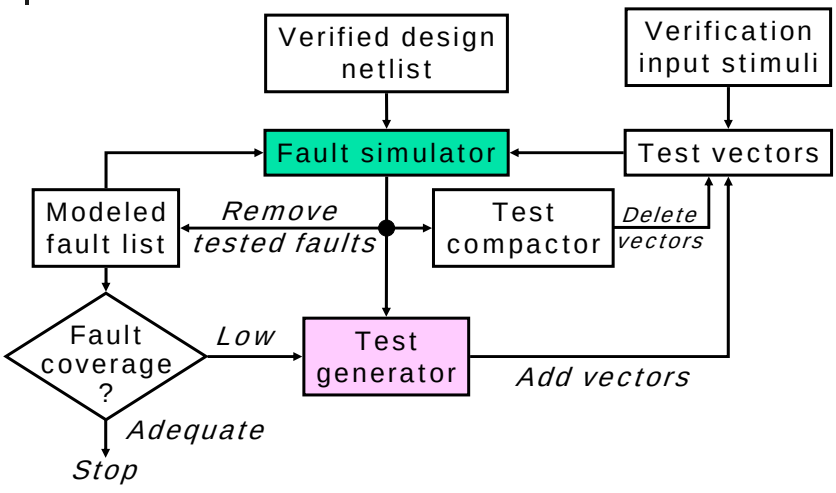
Transition Fault

- two input combinations are needed
  - initialization pattern
    - places an initial value at the fault site.
      - The initial value is 0 for a slow-to-rise transition fault,
      - and 1 for a slow-to-fall transition fault.
  - transition propagation pattern
    - places the final transition value
      - 1 for a slow-to-rise transition fault, and 0 for a slow-to-fall transition fault
    - propagates the transition to an observable output

Fault Simulation

- Fault Simulator
  - A program that models a design with fault present
- Inputs:
  - A circuit
  - A sequence of test vectors
  - A fault model
    - Usually single-stuck faults
    - Sometimes multiple-stuck, bridging faults,...
- Determines
  - Fault coverage (fault grading)
  - Set of undetected faults (Areas of Low Fault Coverage)
  - Generates fault dictionary (Fault diagnosis)
  - Test compaction

Fault simulator in VLSI Design

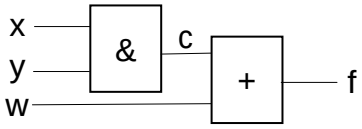


Specific-Fault Oriented Test Generation

- Two fundamental test generation steps
  - **ACTIVATE**, Excite, Provoke or Setup the Fault
    - Make Fault OBSERVABLE, Fault Sensitization
    - Find Primary Input Values that Cause
      - Error Signal in Faulty Circuit
    - For Single-Stuck-at-v Fault
      - Place v' at Fault Site
  - **PROPAGATE** the Resulting Error to a Primary Output
    - Path Sensitization
    - Find Primary Input Values that Sensitize
      - Error Signal to Primary Output

Specific-Fault Oriented Test Generation

- Example: Test for c/0 is w,x,y = 0,1,1
  - ACTIVATE Fault c/0
    - Set x = y = 1 to make c=1
      - in Fault-free Circuit
  - PROPAGATE Value on c to f
    - Set w =0 to sensitize c to f



Test Generation Using Path Tracing

- Notation
  - D Signal Value
    - 1 in Fault-free Circuit, 0 in Faulty Circuit
  - D' or $\overline{D}$ Signal Value
    - 0 in Fault-free Circuit, 1 in Faulty Circuit
  - X
    - Signal Value is Unspecified
- Truth table for AND

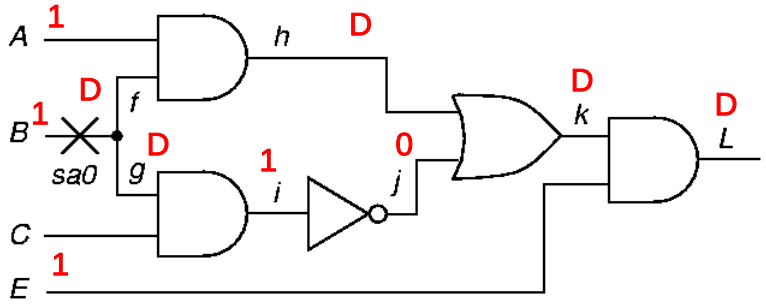
a \ b	0	1	X	D	$\overline{D}$
0	0	0	0	0	0
1	0	1	X	D	$\overline{D}$
X	0	X	X	X	X
D	0	D	X	D	0
$\overline{D}$	0	$\overline{D}$	X	0	$\overline{D}$

Path Sensitization Method

- Fault Sensitization
  - Force tested node to opposite of fault value
- Fault Propagation (path sensitization)
  - Propagate the effect to one or more POs
- Line Justification
  - Justify internal signal assignments made to activate and sensitize fault
- These three steps may result in conflict
  - Different values are assigned to the same signal
  - Require **backtracking**

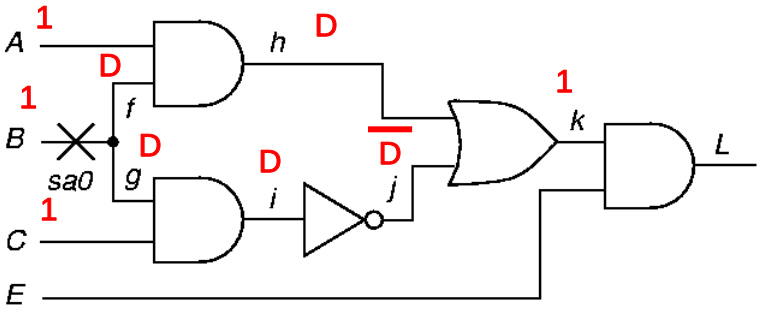
Path Sensitization Method

- Try path f – h – k – L
  - Requires A = 1, j = 0, E = 1
- Blocked at j
  - Since there is no way to justify 1 on i



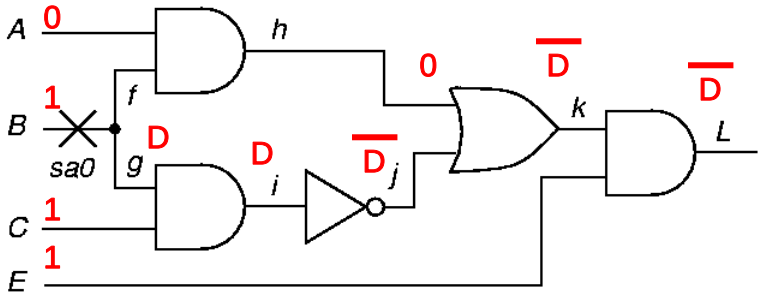
Path Sensitization Method

- Try simultaneous
  - paths f – h – k – L and g – i – j – k – L
- Blocked at k because
  - D-frontier (chain of D or $\overline{D}$) disappears



Path Sensitization Method

- Final try: path g – i – j – k – L
  - test found!



Specific-Fault Oriented Test Generation

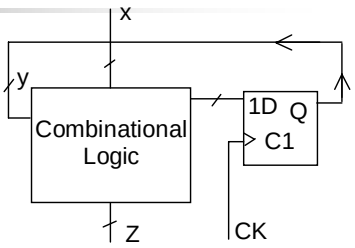
- Three Approaches
  - **D Algorithm**: Internal Line Values Assigned (Roth-1966)
    - D-cubes
    - Bridging faults
    - Logic gate function change faults
  - **PODEM**: Input Values Assigned (Goel – 1981)
    - X-Path-Check
    - Backtracing
  - **FAN**: Input and Internal Values Assigned (1983)

Sequential ATPG

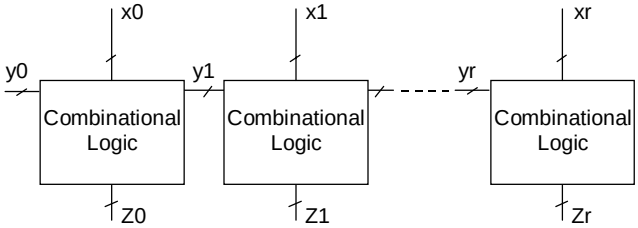
- Approach
  - Convert Finite State Machine to Corresponding Iterative Network
    - Multiple Time Frames (Iterative Cells) Needed for
      - Justification and Propagation
  - One Fault in Sequential Circuit
    - Many Faults in Corresponding Iterative Network
    - Use 9-valued signals
- Issues
  - Order of Justification and Propagation
  - Simulation Values
  - Test Point Insertion (Partial Scan)

General Case

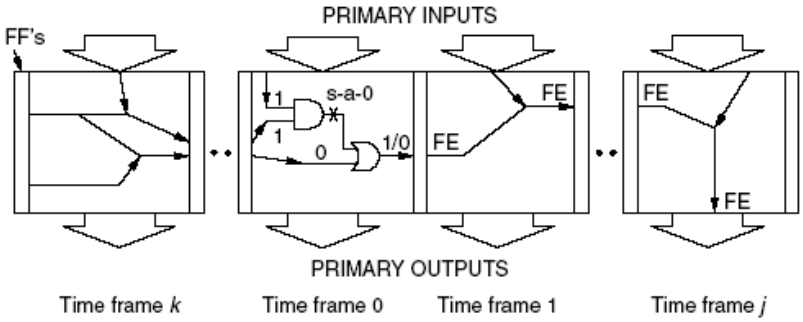
- Huffman model of sequential circuit
  - with edge-triggered D-flip-flops



- Any sequential circuit with edge-triggered D-FF
  - can be directly converted into an iterative network

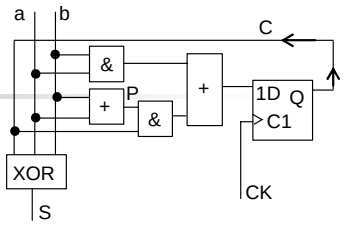


Iterative Logic Array Expansion



- To detect a fault, a sequence of vectors may be needed

Example



- Test for P SA0
  - Provoke Fault on P1: a1 = 0, b1 = 1
  - Propagate Fault to S2:
    - C0 = 1
      - Need to consider last time frame: : a0 = 1, b0 = 1, Cin = X
    - a2 = 0, b2 = 0